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Fault Diagnosis of Marine Diesel Engine Based on Multi-scale Time Domain Decomposition and Convolutional Neural Network

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Abstract

Marine diesel engines work in an environment with multiple excitation sources. Effective feature extraction and fault diagnosis of diesel engine vibration signals have become a hot research topic. Time-domain synchronous averaging (TSA) can effectively handle vibration signals. However, the key phase signal required for TSA is difficult to obtain. During signal processing, it can result in the loss of information on fault features. In addition, frequency multiplication signal waveforms are mixed. To address this problem, a multi-scale time-domain averaging decomposition (MTAD) method is proposed and combined with signal-to-image conversion and a convolutional neural network (CNN), to perform fault diagnosis on a marine diesel engine. Firstly, the vibration signals are decomposed by MTAD. The MTAD method does not require the acquisition of the key phase signal and can effectively overcome signal aliasing. Secondly, the decomposed signal components are converted into 2-D images by signal-to-image conversion. Finally, the 2-D images are input into the CNN for adaptive feature extraction and fault diagnosis. Through experiments, it is verified that the proposed method has certain noise immunity and superiority in marine diesel engine fault diagnosis.

Keywords: multi-scale time-domain averaging decomposition; signal-to-conversion; CNN; marine diesel engine; fault diagnosis

introduction

The diesel engine is the heart of the cabin. It operates in an environment with multiple excitation sources and its vibration signals are characterised by non-linearity and non-stationarity [1]. Research into fault diagnosis is very important, to reduce downtime and improve navigation safety.

In recent years, relevant scholars have analysed vibration signals from the time-frequency domain perspective. The main methods include short-time Fourier transform (STFT) [2], wavelet transforms (WT) [3-5], Hilbert Huang transforms (HHT) [6], and variational mode decomposition (VMD) [7]. Although the above methods are very widely used, they also have certain drawbacks. For example, the window function of the STFT is fixed and is not suitable for the analysis of nonstationary signals [8]. WT needs to manually select the basis function, and different basis functions will affect the diagnosis results [9]. HHT has difficulty in avoiding endpoint effects and modal aliasing during signal decomposition [10,11]. VMD requires the number of decomposition layers and the penalty factor to be set in advance [12]. The difference in parameters will affect the diagnosis results.

Marine diesel engines are rotating machines. Their feature information often concerns the multiples of the rotational frequency. McFadden [13] proposed the time-domain synchronous averaging (TSA) method, which is essentially a series of equidistantly distributed bandpass filters. The

method is effective in suppressing noise interference but has some drawbacks. As a result, it is not widely used in the field of fault diagnosis. Relevant scholars have also suggested ways to improve it. Yin et al. [14] proposed an adaptive multiple TSA method, which solves the period selection problem. Ha et al. [15] used an autocorrelation TSA method to achieve fault diagnosis in wind turbines. Niaki et al. [16] combined VMD with TSA and successfully applied it to the fault diagnosis of the helical gearbox, solving the problem of diagnosing low-intensity faults. In [17], a moving interpolation method was used to optimise the TSA, which improves feature extraction under noisy conditions. In [18], the angular domain window synchronous averaging method was proposed, which overcame the waveform distortion caused by speed variations.

In summary, relevant scholars have improved TSA from different perspectives and successfully applied it in practical engineering. However, there are still certain problems:

- (1)The TSA method tends to overlap specific frequency and frequency multiplication signal waveforms, making it difficult to decompose them effectively.
- (2)After the signal has been averaged in the time domain, it is only possible to obtain feature information at a specific frequency; it is not possible to extract feature information in the full frequency domain at once.
- (3)There is a need to acquire key phase signals. Although sensor technology is developing rapidly, the complex structure of marine diesel engines and the harsh operating environment makes it difficult to obtain key signals.

With advances in intelligence, deep learning methods have been extensively researched. The main methods are CNN, stacked auto-encoding networks, and deep belief networks [18-20]. In [21], a combination of continuous wavelet transforms (CWT) and CNN was used to achieve fault diagnosis of hydraulic piston pumps. Xu et al. [22] used ensemble empirical mode decomposition (EEMD) to decompose the vibration signal and then input its intrinsic modal functions into the CNN to achieve fault diagnosis of the bearing. Zhang et al. [23] obtained 2-D images by wavelet packet transformation of the original signals and input them packet transformation of the original signals and input them
into ResNet to achieve fault diagnosis of the gearbox. In [24], wher the signal-to-image conversion method was used to convert the original signals into two-dimensional images, which perice were then fed into the CNN for fault diagnosis of rotating machinery. Yan et al. [25] achieved fault diagnosis of a ship inverter by 1-D CNN. Li et al. [26] transformed the original signal into a two-dimensional image by using a signal-toimage conversion method and input it into a lightweight CNN, to realise the fault diagnosis of a marine centrifugal fan. Liu et al. [27] used the Gram angle field to transform the signal into a two-dimensional image and improve the residual network, to realise the fault diagnosis of rotor bearings. This shows that deep learning methods have made good progress in the field of fault diagnosis. Deep learning methods can overcome the inherent shortcomings of traditional intelligent fault diagnosis methods, such as the manual extraction of features and the difficulty of learning non-linear relationships. The period of the signal g(t) was $\mathcal{A}(\theta)$ was divided into N segments of length T. The signal g(t) was divided into N segments of length T. $\frac{1}{2}$ iginal can be expressed as: where $\frac{1}{2}$ $\lim_{n \to \infty} \lim_{n \to \infty} \lim_{n \to \infty}$

Inspired by the above literature, a combination of a multiscale time-domain averaging decomposition (MTAD) method, signal-to-image conversion and CNN is proposed for the fault diagnosis in marine diesel engines. The main contributions are as follows:

- (1)The MTAD method can retain the feature information of the original signal more completely. The mutual interference of specific frequencies and frequency multiplication signals is overcome. There is no need to acquire key phase signals. The shortcomings of traditional TSA are overcome.
- (2)The signal components are converted into 2-D images using signal-to-image conversion. This builds a bridge between computer vision and fault diagnosis.
2-D images are into the CNN model for feature extraction and fault diagnosis.
- (3)2-D images are input into the CNN model for feature 2-D mages are mput mio the CIVI moder for reature
extraction and fault diagnosis. Fault diagnosis accuracy can be improved. \overline{C} is the connection and fault diagnosis. Fault diagnosis in \overline{C}

BASIC THEORY BRIDGE BETWEEN

TIME DOMAIN SYNCHRONOUS AVERAGING (TSA)

Given a vibration signal y(t) containing a periodic signal and other signals, the mean value of y(t) is zero. Assuming that and other signals, the mean value of y(t) is zero. Assuming that it satisfies the Dirichlet condition, y(t) can be expressed as: C_{even} a vibuation signal $v(t)$ containing a novia diagnosis.

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y(t) = g(t) + e(t)
$$
\n(1)

where *g(t)* is a periodic signal and *e(t)* are other signals. belief where $g(t)$ is a periodic signal and $e(t)$ are other signals. $\mathcal{A}_\text{max}(\mathcal{A}_\text{max}(\mathcal{A}_\text{max}(\mathcal{A}_\text{max}(\mathcal{A}_\text{max}(\mathcal{A}_\text{max}(\mathcal{A}_\text{max}(\mathcal{A}_\text{max}(\mathcal{A}_\text{max}(\mathcal{A}_\text{max}(\mathcal{A}_\text{max}(\mathcal{A}_\text{max}(\mathcal{A}_\text{max}(\mathcal{A}_\text{max}(\mathcal{A}_\text{max}(\mathcal{A}_\text{max}(\mathcal{A}_\text{max}(\mathcal{A}_\text{max}(\mathcal{A}_\text{max}(\mathcal{A}_\text{max}$

The period of the signal g(t) was T. The signal g(t) was
 thieve divided into N segments of length T.. The nth segment sequence was defined as $P_n(\varphi)$. By the definition of the periodic
function, it follows that: \int $\ln a$. (D) to function, it follows that: tunction, it follows that: σ etion, it follows that: where $\frac{1}{2}$ is a periodic state signal and $\frac{1}{2}$ and $\frac{1}{2}$ are other signals. $\frac{1}{2}$ is zero. As the Dirichlet condition, $\frac{1}{2}$

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P_n(\varphi) = P_{n+1}(\varphi) \tag{2}
$$

 $\mathcal{L}(\mathcal{L}) = \mathcal{L}(\mathcal{L})$ and the transformation of the set of th

where $n \in [1, N]$; phase $\varphi \in [0, T)$.

 μ [24], where $n \in [1, N]$; phase $\varphi \in [0, T)$.
nvert Partition $y(t)$ into N time segments of equal length, with which period T. $Y_n(\varphi)$ is the nth time segment and $Y_n(\varphi)$ can be
tating expressed as: expressed as: expressed as: $\mathbf{1}$ as:

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Y_n(\varphi) = P_n(\varphi) + e_n(\varphi) \tag{3}
$$

 ∇^N () of ∂ since the mean value of $y(t)$ is zero, $\angle n=1$ $\forall n(\forall t)$. $\frac{1}{2}$ -1 ∇^N e (e) \approx \sim \mathfrak{g} o $\sum N$ \cup , \angle ∑() \overline{u} ∇N Ω Since the mean value of y(t) is zero, $\sum_{n=1}^{N} e_n(\varphi) \approx 0$. Since the mean value of $y(t)$ is zero, Since the mean value of y(t) is zero, $\sum_{n=1}^{N} e_n(\varphi) \approx 0$. an value of y(t) is zero, $\angle_{n=1} e_n(\varphi) \approx 0$.

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$$
\sum_{\eta=1}^{N} Y_n(\varphi) \approx \sum_{\eta=1}^{N} P_n(\varphi)
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\n(4)

Based on the above characteristics of $y(t)$, time-domain ships. synchronous averaging can be expressed as the convolution
of $v(t)$ with the periodic signal $g(t)$ [13]: of $y(t)$ with the periodic signal $g(t)$ [13]: Based on the above characteristics of $y(t)$, time-domain
synchronous averaging can be expressed as the convolution $\sum_{i=1}^N$ with the periodic signal $\sum_{i=1}^N$ is less respectively. $\mathcal{L}(\mathcal{C}) = \mathcal{C}(\mathcal{C})$

$$
a(t) = (y * g)(t) = \frac{1}{N} \sum_{n=0}^{N-1} y(t + nT)
$$
 (5) where

$$
x_{1s}^{N}
$$

where $t \in [0, T)$ and $a(t)$ is the TSA result for $y(t)$.

of rotating machinery and effectively suppresses noise be expressed as $Y_r = Y - Y_{fs}$. interference. It is worth noting that the TSA is less resistant to (2) Increase the number of iterations are interference when N is appell. When enclusing less frequency to (4) until the number of iterations are signals, a longer time series is desirable, to prevent the *N* value condition. the prevent increases to obtain the key prevent the key prevent the rotation.
The time domain average decomposite for the rotation of the rotation of the rotation of the rotation of the ro interference when *N* is small. When analysing low-frequency to (4) until from being too small and making it difficult to achieve the (3) The time domain average decondesimed relation can also be agreed as a series of the agreement of the agreement of the agreement of the agreement of the agreem desired result.

desired result.
The prerequisite for the above method is to obtain the
 key phase signals of the rotating machine. Depending on $Y = Y_r + \Sigma Y_{fs}$ (8) the characteristics of the comb filter, the relevant multiplier information can also be fused and difficult to decompose. During the decomposition process m fotating machinery, signals from other periods may also
contain fault information. TSA may also ignore other useful In rotating machinery, signals from other periods may also information when suppressing noise interference. Based on the above analysis, a multi-scale time-domain averaging decomposition method was proposed.

MULTI-SCALE TIME DOMAIN SYNCHRONOUS MULTI-SCALE AVERAGING (MTAD) **MULTIME DOMAIN SYNCHALE THE DOMAIN SYNCHALE THE DOMAIN SYNCHALE THE DOMAIN SYNCHALE**

The MTAD extracts the signal waveform at each frequency, by reconstructing the components of
clique it with the original signal and decomposes it. This hands. For example, the reconstructed aligns it with the original signal and decomposes it. This bands. I method can analyse the features of the signal at different band $[f_s, f_s]$ is defined as: $\frac{1}{2}$ frequencies and preserve the feature frequencies of the original frequencies and preserve the feature frequencies of the original signal more completely. In addition, it can suppress the $V = \sum_{k=1}^{f_{s2}} V f_k$ coherence effects of frequency multiplication signals. It should $Y' = \sum Y f_s$ be noted that the maximum frequency of the decomposition $I = \sum_{f_s = f_{s1}} I_s$ $f_s = f_{s1}$
is determined by the type of fault. The maximum frequency should be greater than the frequency at which the fault is With MTAD, information is determined by the type of fault. The maximum frequency likely to exist. If the fault frequency cannot be determined, frequencies without a key the highest frequency can generally be taken to be 0.5 times subsequent signal process the ingliest request y can generally be taken to be 0.5 times
the sampling frequency. The calculation steps are as follows: (1) The signal duration is t. $f_s = 0.5F_s - i \times s$, Time domain SIGNAL-TO-IMAGE CO ie type of fault. The maximum frequency
than the frequency at which the fault is With MTAD, information can be extracted at different t. $f_s = 0.5F_s - i \times s$, Time domain SIGNAL-TO-IMAGE CONVERSION should be greater than the frequency at which the fault is With MTAD, information can be explicitly the frequency at which the fault is the highest frequency can generally be taken to be 0.5 times subsequent signal proces

synchronous averaging or signals with period $T = \frac{T_s}{f_s}$. r_s is
the sampling frequency. *s* is the step length. In general, \qquad 1-D signals can be con (1) The signal duration is t. $f_s = 0.5r_s = t \times s$, The domain
synchronous averaging of signals with period $T = \frac{1}{f_s}$. F_s is $\frac{f_s}{g}$. F_s is the sampling frequency. *s* is the step length. In general, J. $\frac{a}{4}$ sy the sampling frequency. *s* is the step length. In general, $\frac{200}{F}$ \overline{a}

(2) Dividing the original signal Y into K k^2 of iterations, $i < \frac{r_s}{2s}$, to-imagnoise to-imagnoise the second value of $\frac{r_s}{2s}$ is the second value of $\frac{r_s}{2s}$ in Fig. $S = \frac{200}{tr}$, *i* is the number of iterations, $i < \frac{F_s}{2s}$. (2) Dividing the original signal *Y* into *K* segments by period *T* can be expressed as: . *i* is the number of iterations, $S = \frac{200}{tr}$, *i* is the number of iterations, $i < \frac{F_s}{2}$. period T can be expressed as:
 $L(i)$, $i=1,...,M$ and the pixel (2)Dividing the original signal Y into K^3 segments by $S = \frac{200}{tr_s}$, *i* is the number of iterations,
(2) Dividing the original signal Y in ıs, _i
int $S = \frac{F_s}{tfs}$. It is the number of herations, $t < \frac{F_s}{s}$.
(2) Dividing the original signal Y into K³ segments by $S = \frac{200}{t_{fs}}$, *i* is the number of iterations, $i < \frac{rs}{2s}$.
(2) Dividing the original signal Y into K segn .
gm

$$
Y = \bigcup_{k=1}^{k} \{y_k\} = \bigcup_{k=1}^{k} \{y_t\}, t \in [(k-1)T, kT) \text{ (6)}
$$

$$
P(j,k) = round \{\frac{L((j-1) \times M + k) - Min(L)}{M, k = 1, ..., M} \times 255\} \text{ (1)}
$$

I is a discrete signal that can be interpolated with three times the spline at the corresponding coordinates of $y(t)$. X u reconstruction matrix. $P(j,k) = round$ {
Y is a discrete signal that can be interpolated with three . \mathbf{I} ∞ .
CC \mathbf{c} .
ti onding c . . ιć \mathbf{I} $\boldsymbol{\beta}$.
CC \mathbf{C} $\frac{1}{2}$ onding c .. times the spline at the corresponding coordinates of *y(t)*. *X* is the $\bigcup_{k=1}^{k} \{y_k\}$ -based reconstruction matrix: is the *Y* is a discrete signal that can be interpolated with three times the spline at the corresponding coordinates of $y(t)$. X
is the $\bigcup_{k=1}^{k} \{y_k\}$ -based reconstruction matrix:
where *round*(.) is the round function. Pixel values are

$$
X = \begin{bmatrix} y_1 \\ \cdot \\ \cdot \\ y_K \end{bmatrix} = \begin{bmatrix} x_{11} \dots x_{1n} \\ \cdot \\ \cdot \\ x_{k_1} \dots x_{kn} \end{bmatrix}
$$
normalised from 0 to 2
any manual setting of the
signal can be fully chara
(7)

 $\frac{1}{r}$ where n is the length of each periodic sequence, $n = \frac{T}{F_s}$. .
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where n is the length of each periodic sequence, $n = \frac{T}{F_s}$.

here n is the length of each periodic sequence, $n = \frac{1}{F_c}$.
If we assume that the mean of each column in the matrix X is, then the sequence of periods after averaging over the X \blacksquare $\overline{x} = [\overline{x}_1, \overline{x}_2 ... \overline{x}_n].$ $\overline{X} = [\overline{x}, \overline{x} ... \overline{x}].$ $1, 1, 2, ...$

- TSA extracts the fundamental and octave characteristics times spline interpolation of \overline{X} . The retime domain is $x = [x_1, x_2 ... x_n]$.
 $x = [x, x ... x]$.

(1) The discrete sequence Y_f can be obtained by a three

times spline interpolation of \overline{Y} . The residual signal can ve characteristics times spline interpolation of \overline{X} . The residual signal can
suppresses noise be expressed as $V = V - V$. presses noise be expressed as $Y_r = Y - Y_{fs}$.
	- y suppresses noise be expressed as $Y_r = Y Y_{fs}$.
SA is less resistant to (2)Increase the number of iterations and repeat steps (1)
wsing low-frequency to (4) until the number of iterations exceeds the agreed represent the Nyalue steps (4) until the number of iterations exceeds the agreed
event the Nyalue scondition condition. $\frac{1}{2}$ is (1) and the number of iterations encodes the agreed steps (1) to (1) the number of iterations encodes the agreed
	- o prevent the *N* value condition.
fficult to achieve the (3) The time domain average decomposition result can be expressed as: (3) The time domain average decomposition result can be expressed as: \mathcal{L}_{L} the deterministic splitters splitters splitters splitters splitters. $\exp\left(\frac{\log n}{n}\right)$

$$
Y = Y_r + \Sigma Y_{fs} \tag{8}
$$

er, the relevant multiplier
d difficult to decompose. During the decomposition process, a three-times m other periods may also spline interpolation of the signal is performed for each se interference. Based on when TSA processes discrete signals and can have an impact time-domain averaging on the results. The MTAD uses the amplitude-frequency response feature to separate signals of different frequencies, y also ignore other useful decomposition. This is due to the phase errors that accumulate signals and avoiding the problem of infixing specific requency signals and frequency multiplication signals in TSA. Each component avoiding the problem of mixing specific frequency signals and represents the signal feature of the corresponding frequency. The feature containing the frequency band can be obtained by reconstructing the components of different frequency is bands. For example, the reconstructed signal for frequency nt band $[f_{s1}, f_{s2}]$ is defined as:
al

$$
Y' = \sum_{f_s = f_{s1}}^{f_{s2}} Yf_s
$$
 (9)
imposition
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 ϵ fault frequency cannot be determined, frequencies without a key phase signal, which is the basis for ϵ and ϵ an rrequencies without a key phase signal, which is tr
subsequent signal processing and fault diagnosis. \overline{L} \sim

$w = \frac{1}{r} E$

(2) Dividing the original signal signal signal *Y* is the original *Y* into *M* to-image conversion [28]. The conversion process is shown 2 in Fig. 1. If the obtained image size is *M×M*, the length of the an signal introduced by the segment of the segment of the segment of the segment of the signal is denoted by $\frac{1}{2}$. The value of the signal is denoted by $M, k=1, ..., M$. The formula can be expressed as: 1-D signals can be converted into 2-D images by signal- $L(i)$, $i=1, ..., M$ and the pixel size is denoted by $P(j, k)$, $j=1, ...,$

$$
P(j,k) = round\left\{\frac{L\left((j-1) \times M + k\right) - Min(L)}{\text{Max}(L) - Min(L)} \times 255\right\} \text{ (10)}
$$

any manual setting or the parameterised. $\frac{\text{d}s}{\text{d}t}$ where *round*(*)* is the round function. Pixel values are Ï any manual setting of the parameters and the features in the normalised from 0 to 255. This method does not require

Fig. 1. The signal-to-image conversion process

EXPERI MENTAL ANALYSIS AND DISCUSSIO N

EXPERIMENTAL EQUIPMENT

The experimental diesel engine model was a MAN B&W 6S35ME-B. The rated power was 3250 kW and the rated speed was 142 r/min. The acceleration sensor model was INV9822, with a sampling frequency of 10.24 kHz. The data acquisition instrument was model INV3062 and the experimental platform is shown in Fig. 2. The marine diesel engine fault settings are given in Table 1. During the experiment, the diesel engine speed was 120 r/min and the power was 1800 kW.

Tab.1. Settings for different fault types

Fig. 2. Experimental platform

FAULT DIAGNOSIS PROCESS

The fault diagnosis flow chart is shown in Fig. 3. The main steps were as follows:

Step 1: Vibration signals of different health states were collected.

Step 2: The vibration signals were decomposed by MTAD and the time-frequency energy matrices obtained.

Step 3: 2-D images obtained by signal-to-image conversion and pseudo-colour coding. Training data was input into CNN and a CNN model was built to determine initial learning parameters.

Step 4: The testing data was input into a trained CNN model for feature extraction and fault diagnosis.

Fig. 3. Fault diagnosis flow chart

SIGNAL ANALYSIS

Normal condition and knocking cylinder failure are analysed as examples. The collected vibration signals were decomposed by MTAD at a step size of 0.5. The spectrum of the signal after MTAD decomposition is shown in Fig. 4. The intrinsic frequency of the diesel engine was 1.89 Hz and the signal cylinder firing frequency of the twostroke diesel engine was equal to the rotation frequency (2.00 Hz). Under normal operating conditions, diesel engine vibration is mainly dominated by intrinsic frequency. When a cylinder knocking fault occurs, there is a significant change in vibration amplitude at both the intrinsic frequency and the six-fold frequency. The diesel engine was a sixcylinder diesel engine, so there was a significant change in amplitude at the six-fold frequency. The failure broke the dynamic balance of the engine. The amplitude also became prominent at the two and four-fold frequencies. The problem of waveform mixing between the intrinsic frequency and frequency multiplication signals was overcome by MTAD decomposition.

To link the computer vision and fault diagnosis, 1-D signals were transformed into 2-D images. In this way, invisible information from the original signal could be portrayed. The 1-D feature information was represented by textural features, such as colours and points of the 2-D image. Fig. 5 shows the results of the signal-to-image conversion for the four health states. When the cylinder knocking fault occurs, the time-frequency energy surges and has a certain time interval. For example, the energy intensity is higher at 0.0685 s, 0.2325 s, 0.3941 s, and 0.5592 s. The time interval is about 0.164 s. A lower energy component exists between each of the two higher energies, i.e. the energy intensity changes every 0.082 s. During this period, the frequency is approximately 12.19 Hz, which is very close to the frequency of a cylinder knocking failure (12.27 Hz).

Fig. 4. Spectrogram after MTAD decomposition

Fig. 5. Signal-to-image conversion results for four different health states

CNN MODEL

The CNN model required some hyperparameters to be set before it could be trained and tested. MaxEpoch and Learning rate were used as an example, to demonstrate the hyperparameter determination process. The optional values of MaxEpoch were set as 20, 30, 40, 50, and 80. The optional values of the learning rate was 0.0001, 0.001, and 0.01. The optional value of mini batch size was 50, 75, 100, and 125. The effect of different hyperparameters on the network performance is shown in Figs 6, 7 and 8, respectively. When MaxEpoch = 20, the network model was less time-consuming and had high diagnostic accuracy. Therefore, MaxEpoch was set the value of 20. The network performance is optimal when the learning rate is 0.001. The network performs best when the mini batch size is 50. Other hyperparameters were set in a similar way and the finally determined hyperparameters are shown in Table 2. The CNN model used in this paper had 17 layers, structured as shown in Fig. 9. The specific network model parameters are presented in Table 3.

Fig. 6. Experimental results for different values of MaxEpoch

Fig. 7. Experimental results for different learning rates

Fig. 8. Experimental results for different mini batch sizes

Tab. 2. Results of initial learning parameter settings

LearnRate	0.001
LearnRateDropFactor	0.05
LearnRateDropPeriod	10
L2Regularisation	0.004
MaxEpoch	20
MiniBatchSize	50
Solver	Adam

Fig. 9. The CNN model structure diagram

Tab. 3. The CNN structural parameters

FAULT DIAGNOSIS

The operating environment was Windows 11 and the software environment comprised Anaconda 3, Python 3.9.13 and MATLAB 2021a. The deep learning framework was PyTorch 1.11.0. After pre-training the model with training data, the test data was input into the model for fault diagnosis. The network training process and testing process are shown in Fig. 10. The number of iterations in each round was 98, with a total of 20 iterations, and the maximum number of iterations was 1960. According to Fig. 10(a), as the number of iterations increased, the accuracy of the training data increased until it reached 100%. At the same time, the accuracy rate of the test data also increased and gradually approached 100%. According to Fig. 10(b), the loss function values for both the training and test data gradually decreased until they levelled off. The final value of the loss function for the training data was 0.039. The final value of the loss function for the test data was 0.085. The training time was 251 s and the testing time was 36 s. The CNN model had a superior response speed and no over-fitting of the model occurred during the diagnosis.

Fig. 10. The CNN model training and testing process

In order to validate the CNN model's capability for feature extraction, the clustering effect was visualised and analysed at different stages. The feature distribution of the network structure is analyzed by T-SNE visualization [29]. It can map high-dimensional data to 2-D space well and the visualisation results of different layers are shown in Fig. 11. Different feature information is mixed in the input layer. After two levels of convolution, the same feature information tends to be clustered and different feature information is distributed in different regions. In the softmax layer, the different feature information was completely separated. The network model was very effective in classifying four different health states.

Fig. 11. Results of feature visualisation for different layers

COMPARATIVE ANALYSIS OF DIFFERENT METHODS

The comparative analysis methods used in the experiments were EEMD-CNN [22], MVMD-CNN [12], CWT-CNN [21], and ResNet [23]. In MVMD-CNN, the number of VMD decomposition layers is determined adaptively, using the partial mean of the multi-scale permutation entropy. It can effectively overcome mode aliasing. In CWT-CNN, the original signals are transformed into 2-D images by continuous wavelet transform, which has good timefrequency aggregation. In ResNet, the introduction of residual connectivity enables the extraction of deep feature information while preserving shallow feature information. Different diagnostic methods are analysed in the case of different signal-to-noise ratios (SNR). The diagnostic accuracy of the above methods, at different SNR(s), is shown in Fig. 12.

Fig. 12 shows that, as the noise intensity increased (SNR decreased), the accuracy of the different methods decreased. When SNR = -5 dB, the proposed method achieved a diagnostic accuracy of 94.5%, which is 3.8~8.7% higher than with other methods. When SNR = 30 dB, the confusion matrix of the testing data is shown in Fig. 13. A total of 15 test samples are misclassified. The diagnostic accuracy for the four health states is 99.0%, 99.0%, 98.7% and 98.3%, respectively. The average diagnostic accuracy is 98.8%. In summary, the proposed method is effective in extracting feature information from the original signal and has a certain anti-noise capability.

Fig. 12. Comparison of the diagnostic accuracy of different methods

	$\mathbf i$	297 24.7%	$\overline{3}$ 0.3%	$\mathbf{1}$ 0.1%	$\overline{2}$ 0.2%	98.2% 1.8%
Output Class	$\overline{2}$	$\boldsymbol{0}$ 0.0%	297 24.7%	\mathbf{I} 0.1%	$\mathbf{1}$ 0.1%	99.3% 0.7%
	3	0.1%	$\mathbf{0}$ 0.0%	296 24.6%	$\overline{\mathbf{2}}$ 0.296	99.0% 1.09 _b
	$\frac{4}{1}$	$\mathbf{2}$ $0,2\%$	$\mathbf{0}$ 0.0%	$\overline{2}$ 0.2%	295 24.5%	98.7% 1.3%
		99.0% 1.0%	99,0% 1.0%	98,7% I.3%	98.3% 1.7%	98.8% 1.2%
		\checkmark	γ	J.	\mathbf{v}	
Target Class						

Fig. 13. Confusion matrix for testing data (SNR = 30 dB)

CONCLUSIO NS

In this work, a fault diagnosis method for a marine diesel engine is proposed and validated, based on multi-scale timedomain averaging decomposition, signal-to-image conversion and CNN. The conclusions are as follows:

- (1)MTAD can effectively separate frequency multiplication signals. There is no loss of feature information and there is no need to obtain key phase signals. The effectiveness of MTAD is verified by the experiments. MTAD overcomes the limitations of traditional methods.
- (2)The signal-to-image conversion method can portray the invisible information of the raw signals. The textural features of the 2-D images can make the differences in signal features between working conditions more apparent.
- (3)The accuracy of the proposed method is stable: above 94.0% at different SNR(s). When SNR = -5 dB, the accuracy of the proposed method is 3.818.7% higher than other methods. It has a certain level of adaptive noise immunity.

Due to the limitations of the experimental conditions, only a few typical faults of diesel engines are preset in the experiments. In the next step, a more comprehensive study should be carried out for each subsystem of the diesel engine.

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