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Hardware-efficient algorithms for implementation of the GHM discrete multiwavelet transform kernels

Abstract

In this correspondence, we discuss two efficient algorithms for the execution of forward (FDMWT) and inverse (IDMWT) discrete multiwavelet transform basic operations with reduced computational complexities. We used multiwavelet basis proposed by Geronimo, Hadrin, and Massopust (GHM). The direct implementation of GHM-FDMWT basic operation requires 23 multiplications and 19 additions. The direct implementation of GHM-IDMWT basic operation requires 23 multiplications and 19 additions allow designing the computation procedures, which take only 10 multiplications plus 15 additions for GHM-FDMWT basic operation and 10 multiplications plus 10 additions for GHM-IDMWT basic operation.

Keywords: multiwavelets, GHM, fast algorithms, implementation complexity reduction, FPGA implementation.

1. Introduction

Recently, discrete wavelets transform (DWT) [1] and discrete multiwavelet transform (DMWT) [2] have been used in numerous signal and image processing applications [3-13]. The multiwavelet transform is a new concept in the framework of wavelet transform but it has some important differences. Whereas the wavelet transform has one scaling function and a wavelet function, the multiwavelet transform has two or more scaling functions and mother wavelets for data representation. Multiwavelet transforms possess more superior properties than any single wavelet function such as short support, symmetry, vanishing moments and smoothness. Due to these important properties, they can be used for advanced signal and image processing. Different variations of multiwavelet bases have been developed and presented in the scientific literature. One of the well-known multiwavelet basis obtained with the help of fractal interpolation has been proposed by Geronimo, Hardin, and Massopust (GHM) [10]. GHM basis has become a popular tool for data processing in many computer science applications. Unfortunately, the GHM multiwavelet transform typically requires a significant computational effort [11-14]. This problem becomes extremely challenging for applications requiring real-time processing at high throughput especially in digital signal and image processing. Hence, to meet the stringent requirements to throughput, latency, and power-consumption constraints of real-time DSP systems, the development of dedicated hardware implementations, such as application specific integrated circuits (ASICs) or field-programmable gate arrays (FPGAs), is of paramount importance.

2. Statement of the problem

The "basic operations" of the forward and inverse discrete multiwavelet transform (similar to a "butterfly" operation in the FFT algorithm) are the GHM-FDMWT and GHM-IDMWT kernels, respectively.

In the matrix-vector form, the GHM-FDMWT basic operation can be defined in the following way:

$$\mathbf{Y}_{4\times 1}^{(f)} = \mathbf{F}_{4\times 8} \mathbf{X}_{8\times 1}^{(f)}, \qquad (1)$$

while the inverse operation used to reconstruct signal from GHM-DMWT coefficients (inverse GHM-DMWT base operation) is as follows:

$$\mathbf{X}_{8\times1}^{(i)} = \mathbf{F}_{4\times8}^{(i)} \mathbf{Y}_{4\times1}^{(i)}$$
(2)

where $\mathbf{X}_{8\times 1}^{(f)} = [x_0^{(f)}, x_1^{(f)}, ..., x_7^{(f)}]^T$ - is an 8-element input data vector and $\mathbf{Y}_{4\times 1}^{(f)} = [y_0^{(f)}, y_1^{(f)}, y_2^{(f)}, y_3^{(f)}]^T$ - is a 4-element output data vector for GHM-FDMWT kernel (1), $\mathbf{Y}_{4\times 1}^{(i)} = [y_0^{(i)}, y_1^{(i)}, y_2^{(i)}, y_3^{(i)}]^T$ - is an 4-element input data vector and $\mathbf{X}_{8\times 1}^{(i)} = [x_0^{(i)}, x_1^{(i)}, ..., x_7^{(i)}]^T$ - is a 8-element output data vector for GHM-IDMWT kernel (2), and superscripts *f* and *i* indicate the forward and inverse transforms, respectively.

Matrix
$$\mathbf{F}_{4\times8}^{(f)} = \begin{bmatrix} \mathbf{H}_{2}^{(0)} \mid \mathbf{H}_{2}^{(1)} \mid \mathbf{H}_{2}^{(2)} \mid \mathbf{H}_{2}^{(2)} \mid \mathbf{H}_{2}^{(3)} \\ \mathbf{G}_{2}^{(0)} \mid \mathbf{G}_{2}^{(1)} \mid \mathbf{G}_{2}^{(2)} \mid \mathbf{G}_{2}^{(3)} \end{bmatrix}$$
 (3)

in (1) is a GHM-FDMWT base matrix with dimensions of (4×8) whose submatrices $\mathbf{H}_2^{(l)}$ and $\mathbf{G}_2^{(l)}$ contain the coefficients of the high-pass and low-pass matrix-valued filter impulse responses, respectively.

In turn, the matrix:

$$\mathbf{F}_{8\times4}^{(i)} = \begin{bmatrix} \mathbf{[H}_{2}^{(0)}]^{\mathrm{T}} & [\mathbf{G}_{2}^{(0)}]^{\mathrm{T}} \\ \mathbf{[H}_{2}^{(1)}]^{\mathrm{T}} & [\mathbf{G}_{2}^{(1)}]^{\mathrm{T}} \\ \mathbf{[H}_{2}^{(2)}]^{\mathrm{T}} & [\mathbf{G}_{2}^{(2)}]^{\mathrm{T}} \\ \mathbf{[H}_{2}^{(3)}]^{\mathrm{T}} & [\mathbf{G}_{2}^{(2)}]^{\mathrm{T}} \\ \mathbf{[H}_{2}^{(3)}]^{\mathrm{T}} & [\mathbf{G}_{2}^{(3)}]^{\mathrm{T}} \end{bmatrix}$$
(4)

in (2) is a GHM-IDMWT base matrix with dimensions of (8×4), and submatrices $[\mathbf{H}_2^{(l)}]^{\mathrm{T}}$ and $[\mathbf{G}_2^{(l)}]^{\mathrm{T}}$ are the transposed matrices of the matrices $\mathbf{H}_2^{(l)}$ and $\mathbf{G}_2^{(l)}$.

For the GHM basis these submatrices are as follows:

$$\mathbf{H}_{2}^{(0)} = \begin{bmatrix} \frac{3}{5}\sqrt{2} & | & \frac{4}{5} \\ -\frac{1}{20} & | & -\frac{3}{10}\sqrt{2} \end{bmatrix}, \ \mathbf{H}_{2}^{(1)} = \begin{bmatrix} \frac{3}{5}\sqrt{2} & | & 0 \\ 9/20 & | & 1/\sqrt{2} \end{bmatrix},$$
$$\mathbf{H}_{2}^{(2)} = \begin{bmatrix} \frac{0}{9/20} & | & -\frac{3}{10}\sqrt{2} \\ \frac{9}{20} & | & -\frac{3}{10}\sqrt{2} \end{bmatrix}, \ \mathbf{H}_{2}^{(3)} = \begin{bmatrix} \frac{0}{-1/20} & | & 0 \\ -\frac{1}{20} & | & 0 \\ \frac{1}{10}\sqrt{2} & | & -\frac{3}{10}\sqrt{2} \end{bmatrix}, \ \mathbf{G}_{2}^{(1)} = \begin{bmatrix} \frac{9}{20} & | & -\frac{1}{\sqrt{2}} \\ -\frac{9}{10}\sqrt{2} & | & -\frac{1}{\sqrt{2}} \end{bmatrix},$$
$$\mathbf{G}_{2}^{(2)} = \begin{bmatrix} \frac{9}{20} & | & -\frac{3}{10}\sqrt{2} \\ \frac{9}{10}\sqrt{2} & | & -\frac{3}{10}\sqrt{2} \end{bmatrix}, \ \mathbf{G}_{2}^{(3)} = \begin{bmatrix} -\frac{1}{20} & | & 0 \\ -\frac{1}{10}\sqrt{2} & | & 0 \\ -\frac{1}{10}\sqrt{2} & | & 0 \end{bmatrix}.$$

Formulas (1), (2) define the GHM-FDMWT and GHM-IDMWT basic kernels, respectively.

Substituting the numerical values of $\mathbf{H}_{2}^{(l)}$ and $\mathbf{G}_{2}^{(l)}$ into the formula (3) and numerical values of $[\mathbf{H}_{2}^{(l)}]^{\mathrm{T}}$ and $[\mathbf{G}_{2}^{(l)}]^{\mathrm{T}}$ into the formula (4) we obtain:

$$\mathbf{F}_{4\times8} = [\mathbf{F}_4^{(0)} \mid \mathbf{F}_4^{(1)}] \tag{5}$$

$$\mathbf{F}_{4}^{(0)} = \begin{bmatrix} \frac{3/5\sqrt{2}}{-1/20} & \frac{4/5}{-3/10\sqrt{2}} & \frac{3/5\sqrt{2}}{9/20} & \frac{0}{1/\sqrt{2}} \\ \frac{-1/20}{-1/20} & \frac{-3/10\sqrt{2}}{9/20} & \frac{9/20}{-1/\sqrt{2}} \\ \frac{-1/20}{1/10\sqrt{2}} & \frac{3/10\sqrt{2}}{3/10} & \frac{-9/20}{-9/10\sqrt{2}} & \frac{-1/\sqrt{2}}{0} \end{bmatrix}$$
(6)
$$\mathbf{F}_{4}^{(1)} = \begin{bmatrix} \frac{0}{9/20} & \frac{0}{-3/10\sqrt{2}} & \frac{0}{-1/20} & \frac{0}{0} \\ \frac{9/20}{-3/10\sqrt{2}} & \frac{-1/20}{-1/10\sqrt{2}} & \frac{0}{0} \\ \frac{9/20}{9/10\sqrt{2}} & \frac{-3/10\sqrt{2}}{-3/10} & \frac{-1/20}{-1/10\sqrt{2}} & \frac{0}{0} \end{bmatrix}$$
(6)

$$\mathbf{F}_{8\times4} = \begin{bmatrix} \frac{3/5\sqrt{2}}{4/5} & -1/20 & -1/20 & 1/10\sqrt{2} \\ \frac{4/5}{4/5} & -3/10\sqrt{2} & -3/10\sqrt{2} & -3/10 \\ \frac{3/5\sqrt{2}}{9/20} & 9/20 & 9/20 & -9/10\sqrt{2} \\ \frac{-0}{0} & 1/\sqrt{2} & -1/\sqrt{2} & 0 \\ \frac{-0}{0} & 9/20 & 9/20 & 9/10\sqrt{2} \\ \frac{-0}{0} & -3/10\sqrt{2} & -3/10\sqrt{2} & -3/10 \\ \frac{-0}{0} & 0 & 0 & 0 \end{bmatrix}$$
(7)

From expressions (1) and (2) it is clear that the implementation of the GHM-FDMWT basic operations required carrying out 23 multiplications and 19 additions whereas the implementation of the GHM-IDMWT basic operations required making 23 multiplications and 16 additions. We show two algorithms that require a much smaller amount of computations for the implementation of the GHM-FDMWT and GHM-IDMWT kernels.

3. The algorithms

The first algorithm (for implementation of the GHM-FDMWT kernel) can be written with the help of the following matrix-vector calculating procedure:

$$\mathbf{Y}_{4\times 1}^{(f)} = \mathbf{W}_{4\times 6} \mathbf{D}_6 \mathbf{W}_{6\times 9} \mathbf{D}_9 \mathbf{W}_{9\times 8} \mathbf{X}_{8\times 1}^{(f)}$$
(3)

where

$$\mathbf{W}_{9\times8} = \begin{bmatrix} 1 & 1 & & & \\ 1 & & & & 1 \\ 1 & & & & 1 \\ 1 & & & -1 & & \\ & -1 & 1 & & & \\ & 1 & & 1 & & \\ 1 & & & -1 & & \\ 1 & & & 1 & & \\ \end{bmatrix}$$
$$\mathbf{D}_{9} = diag(\frac{3}{5\sqrt{2}}, \frac{4}{5}, 1, 1, 9, 9, 1, \frac{3}{10}, \frac{3}{10}),$$
$$\mathbf{D}_{6} = diag(1, \frac{1}{20}, \frac{1}{10\sqrt{2}}, \frac{1}{2}, \frac{1}{2}, 1),$$



The second algorithm (for implementation of the GHM-IDMWT kernel) can be written with the help of the following matrix-vector calculating procedure:

$$\mathbf{X}_{8\times1}^{(i)} = \mathbf{W}_{8\times9}\mathbf{D}_{9}\mathbf{W}_{9\times6}\mathbf{D}_{6}\mathbf{W}_{6\times4}\mathbf{Y}_{4\times1}^{(i)}$$
(4)

where

$$\mathbf{W}_{8\times9} = [\mathbf{W}_{9\times8}]^{\mathrm{T}}, \ \mathbf{W}_{9\times6} = [\mathbf{W}_{6\times9}]^{\mathrm{T}}, \ \mathbf{W}_{6\times4} = [\mathbf{W}_{4\times6}]^{\mathrm{T}}.$$

Fig. 1 shows a data flow diagram of the proposed algorithm for realization of the GHM-FDMWT kernel, whereas Fig. 2 shows a data flow diagram of the proposed algorithm for realization of the GHM-IDMWT kernel. In this paper, the data flow diagrams are oriented from left to right. The straight lines in the figures denote the operations of data transfer. The circles in these figures show the operation of multiplication by a number inscribed inside the circle. The points where the lines converge denote summation the dotted lines indicate the sign-change operations. We use the usual lines without arrows purposefully so as not to clutter the picture.



Fig. 1. The data flow diagram of the proposed algorithm for realization of the GHM-IDMWT kernel

Taking into account the presence of zero elements in the matrices $\mathbf{F}_{4\times8}^{(f)}$ and $\mathbf{F}_{8\times4}^{(i)}$, the actual number of multiplications needed to implement the GHM-FDMWT/GHM-IDMWT kernels is 23 instead 32. To implement our algorithms, it is required to carry out only 10 multiplications. This is more than twice reduction.

Minimizing the number of multiplications is especially important in the design of specialized VLSI on-board DSP processors because reducing the number of multipliers also reduces the power dissipation and lowers the implementation cost of the entire system being implemented. Moreover, a hardware multiplier is a more complicated unit than an adder and occupies much more chip area than the adder. Even if the VLSI chip already contains embedded multipliers, their number is always limited. This means that if the implemented algorithm has a large number of multiplications, the projected processor may not always fit into the chip and the problem of minimizing the number of multiplications remains relevant [15].



Fig. 2. The data flow diagram of the proposed algorithm for realization of the GHM-IDMWT kernel

4. Conclusions

The paper presents two new FPGA-oriented algorithm for the execution of GHM-FDMWT and GHM-IDMWT kernels with reduced computational complexities. To reduce the hardware complexity (number of embedded multipliers), we exploit the specific structural properties of the matrix-vector products that represent GHM-FDMWT and GHM-IDMWT basic operations. So, the fully parallel implementation of the GHM-FDMWT kernel requires 8 multipliers by fractional numbers, two multipliers by integer numbers and 15 adders. The fully parallel implementation of the GHM-IDMWT kernel requires also 8 multipliers by fractional numbers, two multipliers by integer numbers and only 10 adders. For example, a fully parallel implementation of the GHM-FDMWT kernel or the GHM-IDMWT kernel using schoolbook (direct) method requires two FGPA-chips Spartan-3 XC3S200, while the implementation of these kernels with the help of our proposed algorithms occupies only one such chip.

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5. References

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